

Best Available Copy

(12) INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(19) World Intellectual Property Organization
International Bureau



(43) International Publication Date
21 March 2002 (21.03.2002)

(10) International Publication Number
PCT WO 02/23742 A2

(51) International Patent Classification⁷:

H04B

(74) Agent: STEPHENS, Gregory, A.; Gregory A. Stephens, P.C., 301 Edgemore Avenue, Apex, NC 27502 (US).

(21) International Application Number: PCT/US01/28051

(22) International Filing Date:

7 September 2001 (07.09.2001)

(25) Filing Language:

English

(26) Publication Language:

English

(30) Priority Data:

09/660,050 12 September 2000 (12.09.2000) US

(71) Applicant (for all designated States except US): ERICSSON INC. [US/US]; 7001 Development Drive, Research Triangle Park, NC 27709 (US).

(72) Inventors; and

(75) Inventors/Applicants (for US only): HUSEYIN, Arslan [TR/US]; 600 Aaron Circle, #614, Durham, NC 27713 (US). MOLNAR, Karl, James [US/US]; 110 Flying Leaf Court, Cary, NC 27513 (US). BOTTOMLEY, Gregory, E. [US/US]; 100 Merlot Court, Cary, NC 27511 (US). KHAYRALLAH, Ali, S. [LB/US]; 103 Streamview Drive, Apex, NC 27502 (US).

(81) Designated States (national): AE, AG, AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, BZ, CA, CH, CN, CO, CR, CU, CZ, DE, DK, DM, DZ, EE, ES, FI, GB, GD, GE, GH, GM, HR, HU, ID, IL, IN, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MA, MD, MG, MK, MN, MW, MX, MZ, NO, NZ, PL, PT, RO, RU, SD, SE, SG, SI, SK, SL, TJ, TM, TR, TT, TZ, UA, UG, US, UZ, VN, YU, ZA, ZW.

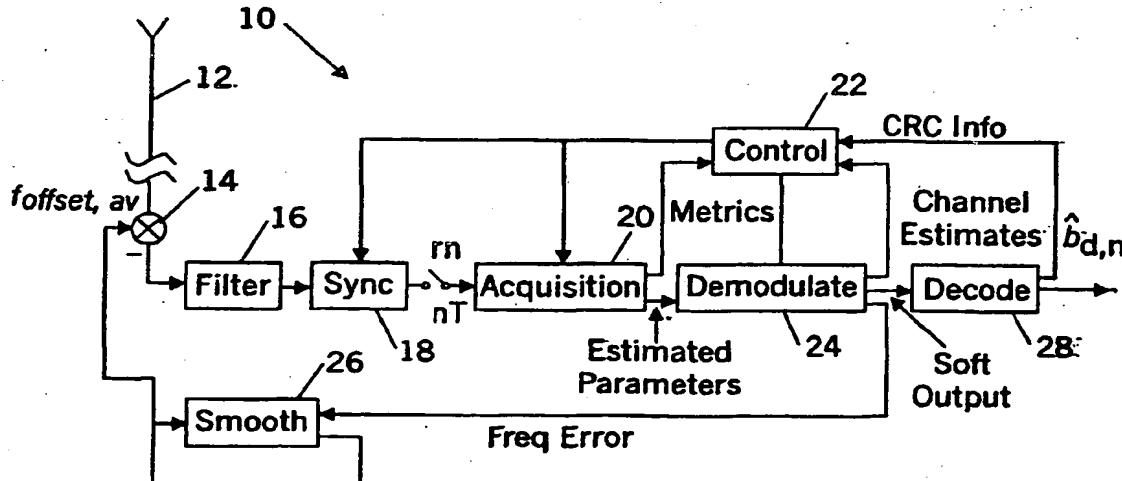
(84) Designated States (regional): ARIPO patent (GH, GM, KE, LS, MW, MZ, SD, SL, SZ, TZ, UG, ZW), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE, TR), OAPI patent (BF, BJ, CF, CG, CI, CM, GA, GN, GQ, GW, ML, MR, NE, SN, TD, TG).

Published:

— without international search report and to be republished upon receipt of that report

For two-letter codes and other abbreviations, refer to the "Guidance Notes on Codes and Abbreviations" appearing at the beginning of each regular issue of the PCT Gazette.

(54) Title: APPARATUS FOR AND METHOD OF ADAPTING A RADIO RECEIVER USING CONTROL FUNCTIONS



WO 02/23742 A2

(57) Abstract: An adaptive radio receiver utilizes control functions derived from received signals in an interference-limited environment to decide whether to implement conventional single-user demodulation or joint demodulation. The decision to implement joint or conventional demodulation is based upon the values of estimates derived for Doppler spread, carrier-to-interference plus noise, dispersiveness and the likelihood of the presence of a dominant interferer.

APPARATUS FOR AND METHOD OF ADAPTING A RADIO RECEIVER USING CONTROL FUNCTIONS

BACKGROUND OF THE INVENTION

The present invention relates to using control functions to adapt a radio receiver to selectively perform either conventional or joint demodulation.

The cellular telephone industry has made phenomenal strides in commercial operations in the United States as well as the rest of the world. Growth in major metropolitan areas has far exceeded expectations. Solutions are required to meet these increasing capacity needs and to maintain high quality service.

As additional users are introduced to wireless cellular systems, the system characteristics change from a noise-limited to an interference-limited environment. Interference problems can arise when signals transmitted between locations suffer from echo distortion or multipath time dispersion caused, for example, by signal reflections from large buildings or nearby mountain ranges. Multipath dispersion occurs when a signal proceeds to the receiver along not one, but many paths, so that at the receiver many echoes having different and randomly varying delays and amplitudes occur. For example, when multi-path time dispersion is present in a TDMA system, the receiver receives a composite signal comprising multiple versions of the transmitted symbol that have propagated along different paths having different relative time delays. At the receiver, such multiple versions of the transmitted signal can cause self-interference.

In addition to being performance limited by self-interference, receiver performance also may be limited by co-channel signal interference. Work has been done

to address the problem of co-channel signal interference by means of joint demodulation.

Note, for example, the discussions of and approaches to joint demodulation that are found in US Patents Nos. 5,506,861, 5,790,549 and 5,790,606, assigned to the assignee of the present invention and the teachings of which are incorporated herein by reference.

SUMMARY OF THE INVENTION

The present invention relates to a radio receiver comprising a Doppler spread estimator for generating an estimated Doppler spread value of the carrier; a first estimator generates an estimated carrier-to-dominant interferer plus noise ratio value, and a second estimator generates an estimated dominant interferer plus noise-to-noise ratio value. In accordance with one aspect of the invention, a control unit receives the estimated Doppler spread value, the estimated carrier-to-dominant interferer plus noise ratio value, and the estimated dominant interferer plus noise-to-noise ratio value, and generates control signals to control the radio receiver to selectively perform conventional or joint demodulation.

addition, a control unit selectively controls the radio receiver, in accordance with the estimated values, to selectively implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer.

The invention also contemplates a method of operating a radio receiver, which receives a signal that includes a desired carrier, a co-channel dominant interferer and noise. The method comprises generating an estimated Doppler spread value of the carrier, developing an estimated carrier-to-dominant interferer plus noise ratio value, and providing an estimated dominant interferer plus noise-to-noise ratio value. In addition, the method controls the radio receiver, in accordance with the estimated values, to selectively implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer.

The foregoing and other advantages and features of the invention will become apparent upon a consideration of the following description, when taken in conjunction with the accompanying drawings.

BRIEF DESCRIPTION OF THE DRAWINGS

Fig. 1 is a block diagram of a control portion of a radio receiver, for selectively controlling the receiver to use conventional or joint demodulation in accordance with the values of control functions derived from received signals;

Fig. 2 is a block diagram of the control unit of Fig. 1;

Fig. 3 is a block diagram of a Doppler spread estimator of a type that may be used by the receiver;

Fig. 4 is a block diagram of the demodulate unit according to the invention;

Figs. 5 and 6 are flow charts showing contemplated modes of operation of the radio receiver; and

Fig. 7 is a generalized illustration of a mobile communications system including a base station and a mobile terminal in accordance with the invention.

DETAILED DESCRIPTION

Co-channel interference is a major impairment to performance of radio receivers, such as mobile communications system radio receivers. It has been found that joint demodulation can improve the performance of a receiver in interference-limited environments, but at the expense of computational complexity. The complexity increases with the number of users and the number of desired signal and interferer signal channel

taps used in the demodulator or equalizer. The present invention contemplates selective use of conventional or joint demodulation in an adaptive radio receiver in interference-limited environments. To limit computational complexity, it is assumed that only one single dominant interferer is present and degrading receiver performance. The presence 5 of a dominant interferer at a receiver is very likely when the receiver experiences bad signal quality, i.e., a low carrier-to-interference ratio C/I . Thus, the invention contemplates identifying, under the condition of bad signal quality at a receiver and for the purpose of adapting the receiver to employ joint demodulation, whether a dominant interferer exists.

10 Although joint demodulation improves receiver performance when a dominant interferer exists, there can be situations where there is no or very low interference, in which using a joint demodulator might degrade receiver performance, while increasing computational complexity unnecessarily. In addition, even if channel quality is low, where conventional demodulation or equalization does not perform well, joint demodulation might not perform well either, such as in the situation where a low C/I is being caused by the presence of more than one interferer at the receiver. Therefore, it is contemplated that the receiver use joint demodulation only when there is a single strong co-channel interference present in the received signal, and that the receiver otherwise use conventional demodulation.

15 20 The invention takes into account that joint demodulation is sensitive to dispersion in the desired channel response, but is not highly sensitive to dispersion in interferer channel response. When channels are dispersive, channel acquisition and channel estimation algorithms do not perform as well as when the channel is flat. If it is known that a channel is flat, receiver performance can be improved by changing the channel 25 acquisition and estimation parameters accordingly, while reducing the required computational complexity considerably. The invention therefore contemplates use of a control function that identifies the dispersiveness of the channel as a factor for consideration in deciding of whether to employ joint demodulation to improve receiver performance.

30 For low-speed mobile receivers, such as in mobile terminals, joint demodulation improves the receiver performance considerably, in the presence of a dominant interferer, compared to conventional demodulation. For moderate-speed mobile receivers, the gains in receiver performance afforded by joint demodulation may be limited. For high-speed mobile receivers, joint demodulation may not perform as well as conventional

demodulation for some 'C/I' levels. As a result, the invention further contemplates identifying the Doppler spread value of the desired signal and using joint demodulation selectively, depending on the Doppler spread information.

The invention makes use of several control functions in an adaptive radio receiver that has the ability of demodulating co-channel signals jointly. With such control functions and the values of the control functions, the adaptive receiver selects the right demodulator or parameters for the signal conditions then existing. As a result, performance of the receiver is improved while computational complexity is minimized.

The contemplated control functions (speed estimation, noise estimation, dominant interference estimation; dispersive or non-dispersive channel identification, etc.) are used by the receiver in the process of making the decisions whether to use joint demodulation or conventional demodulation. The decisions are made on a slot-by-slot basis. The control functions attempt to turn on joint demodulation by the receiver in favorable conditions, falling back to conventional demodulation or equalization in fast fading or dispersive signal conditions. Because conventional demodulation is less complex, multipass demodulation or other enhancements can be used for improving receiver performance.

Fig. 1 shows a block diagram of part of an adaptive radio receiver system 10 in which joint demodulation is integrated with conventional single-user demodulation in the receiver, such that joint demodulation is selectively available to the receiver for processing received signals under favorable conditions as determined by the values of control functions. In the described implementation of the invention, joint demodulation is not preferable to conventional equalization under receiver signal environments where: (1) no or very limited signal interference is present; (2) high mobile speeds exist; (3) the desired signal is heavily dispersive; and/or (4) multiple dominant interferers exist.

Clearly, when no interference is present, it is better to turn joint demodulation off and use conventional demodulation, rather than estimate quantities for an interfering user that is not there. Also, the conventional equalizer is tuned for higher mobile speeds and a dispersive signal, with the consequence that channel estimation becomes the limiting factor when more quantities have to be estimated due to addition of an interfering user, which is especially true for some fixed level of computational complexity. For such reasons, the implementation of the described embodiment of the invention is such as to detect only one dominant interferer as a criteria employing joint demodulation.

Priority is also claimed from U.S. Provisional Application No. 60/218,320, filed July 23, 2000.

The present invention is described herein in the context of a mobile terminal 200 of a mobile communications system 202, see Fig. 7. The receiver system 10 may also be implemented in a base station receiver 204 of a base station 206 of the mobile communications system 202, as will be apparent. As used herein, the term "mobile terminal" may include a mobile communications radiotelephone with or without a multi-line display; a Personal Communications System (PCS) terminal that may combine a mobile communications radiotelephone with data processing, facsimile and data communications capabilities; a PDA that can include a radiotelephone, pager, Internet/intranet access, Web browser, organizer, calendar and/or a global positioning system (GPS) receiver; and a conventional laptop and/or palmtop receiver or other appliance that includes a radiotelephone transceiver. Mobile terminals may also be referred to as "pervasive computing" devices.

The part of the receiver system 10 shown in Fig. 1 includes a receiver front end comprising an antenna 12, a multiplier block 14, and a filter block 16 where the received signal is downconverted, frequency corrected according to an estimated frequency offset of the desired signal, and filtered. Although a single antenna 12 is illustrated, the receiver system 10 could use more than one antenna. A sync block 18 inputs the baseband signal from the filter 16, and selects one of the sampling phases depending on the sync point chosen from correlating to the desired signals sync sequence. Conventional demodulation uses a two-tap sync algorithm which is a sync algorithm designed for dispersive channels. When joint demodulation is considered, a one-tap sync point (designed for flat fading channels) is also computed since only one desired signal tap is assumed for joint demodulation. After acquisition, the synchronization point for the chosen demodulation method is used for further processing. The sync block outputs the received signal samples (τ_n) at sampling time (nT), where n is the sample number and T is the sample period.

An acquisition block 20 performs the function of training the channel for the desired signal, starting with generating initial channel estimates computed from correlations at the selected sync point. Additionally, for joint demodulation, the interferer channel and delay timing is acquired in the acquisition block 20. Different delay timings are hypothesized, and training is performed jointly for both the desired signal and the interferer signal. This may be done as described in an application entitled "Method and Apparatus for Co-channel Interference", serial no. 09/143,821, filed August 31, 1998. A

control block 22 uses metric values for both conventional and joint demodulation to determine which demodulation approach should be used for data detection.

A demodulate block 24 uses either conventional or joint detection to demodulate the unknown data. It outputs hard bit values for uncoded transmitted bits and soft bit values for coded bit data. The demodulation process also outputs channel estimates that are used in the control function performed by the control block 22 to estimate Doppler speed, and frequency estimates that are used by a smooth block 26 to correct the signal clock due to frequency error in the received signal. In addition, the output from the demodulate block 24, which can be generated by either the conventional or joint demodulation process, is input to a decode block 28. The output of the decode block 28 provides cyclic redundancy code (CRC) information to the control block 22, which is used as a measure of signal frame validity. The CRC information also is used in the long-term Doppler estimation process, and indirectly affects which approach is chosen for the next data slot's demodulation.

Frequency error of the received signal with respect to the desired signal is estimated, filtered and used to further correct the received signal in subsequent data slots.

The control block 22 performs the control function and determines whether the estimated conditions are favorable for using joint demodulation, as compared to conventional demodulation. If conditions are not favorable for joint demodulation, the control block 22 selects conventional equalization for use. Favorable conditions can occur, for example, when the mobile is in a slow, flat fading environment when a single dominant interferer is present. The control function must therefore differentiate between slow and fast Doppler speeds, flat and dispersive fading, and the presence and absence of a single dominant interferer.

Fig. 2 shows the control block 22 of Fig. 1, as a control unit which uses information obtained from the acquisition, demodulation and decoding algorithms to decide whether to perform single-user demodulation or joint demodulation. The invention contemplates that the decision be made on the basis of three channel condition indicators: (1) the Doppler spread for the desired signal ($f_{D,ss}$); (2) the carrier-to-interference plus noise ratio $C/(I+N)$ or signal-to-interference plus noise ratio SINR, and (3) the dominant interferer plus noise-to-noise ratio $(I+N)/N$, where here and throughout the disclosure C is a desired signal, I is a co-channel dominant interferer, and N is all other noise, i.e., the sum of all other co-channel interferers plus thermal noise.

These indicators are estimated by using the accumulated metrics obtained from single-user and joint acquisition, the instantaneous channel amplitude estimates obtained from demodulation, and the CRG information obtained from decoding.

One of the major control functions used in determining whether to adapt the receiver 10 to employ joint demodulation is the Doppler spread estimator. As mentioned, joint demodulation does not perform well for high Doppler spread values, receiver gains are limited for intermediate Doppler spread values, and gains in receiver performance are significant for low Doppler spread values. Consequently, the control block 22 uses a Doppler spread estimate in making a decision whether to adapt the receiver to use the joint detector (low Doppler spread values) or to use the conventional detector (high Doppler spread values).

One contemplated, but not exclusive, technique for determining Doppler spread estimate values is as follows:

(1)

$$\hat{D}(n) = |\hat{c}(n)| - |\hat{c}(n-r)|$$

15

where $D(n)$ represents the absolute differentials of the channel tap estimate, and

$\hat{c}(n)$ is the channel estimate at time n . By using absolute differentials, the effect of frequency error is decoupled from Doppler spread estimation. If $r = 1$, since the differentials of the channel estimates are very noisy, low pass filtering (or averaging) of

20 $D(n)$ is required.

(2)

The adaptation set for estimating the Doppler spread includes both the absolute differentials and the low pass filter. The inclusion of both allows for a more accurate estimate of the Doppler spread. The low pass filter is used to smooth the differentials to prevent high noise values. The formula for the low pass filter is:

$$\tilde{D}_k = \frac{1}{N} \sum_{n=1}^N D(n)$$

The bandwidth of the low pass filters, or the averaging window size N , is also a function of the Doppler spread estimate. Additionally, the adaptation parameters of the channel estimator depend on the Doppler spread estimation. Using erroneous adaptation parameters directly affects the performance of the Doppler estimator.

A single receiver antenna and flat fading channel conditions are assumed for Doppler spread estimation. Therefore, the differentials and Doppler estimates are obtained from a single channel tap. However, the same approach can be applied to a

dispersive channel by either considering the strongest channel tap estimates, or combining the results from multiple taps. Similarly, the same approach can be applied for multiple antennas by combining the results from multiple antennas.

Fig. 3 is a block diagram of the Doppler spread estimator 30 of a type as may be embodied in the adaptive receiver 10. The receiver 10 starts with some initial Doppler value and channel tracker parameters 32. Using the channel estimate over the current slot, the Doppler value is updated, and the updated value is then used to adapt the parameters of the Doppler estimator 30 and the channel tracker 34 for the subsequent slots.

To calculate the Doppler spread estimate, begin with the initial Doppler estimate

$$(3) \quad \hat{f}_D = \text{Doppler_init}.$$

The initial channel tracker parameters and Doppler filter parameters are calculated from the initial Doppler estimate 32. For an AR2 type channel tracker 34, there are three parameters in the algorithm: w_d is the Doppler spread that is calculated using the

Doppler estimate as $w_d = \frac{2\pi}{R_s} \hat{f}_D$, where R_s is the symbol rate; λ is a form of adaptation

gain which can be derived as

$$\lambda = 35714 \cdot 10^{-6} * \hat{f}_D + 1 \cdot 10^{-6};$$

and r_d , which determines the height of the spectral peak at w_d , is kept constant with a value of $r_d = 0.998$, in an exemplary embodiment.

If the cyclic redundancy code CRC does not indicate to the contrary, the instantaneous Doppler value over the whole slot is calculated. The slot is divided into sub-blocks of length N . Depending on the block length, each slot will have K sub-blocks. Both N and K are variables that depend on the previously calculated Doppler spread value. The slot consists of 162 channel estimates, $\hat{c}(1), \dots, \hat{c}(162)$. The differentials are calculated from the channel estimates as $\hat{D}(n) = |\hat{c}(n)| - |\hat{c}(n-1)|$, $n=1 \dots 161$. Over each sub-block, the differentials are averaged as

$$\tilde{D}_F = \frac{R_s}{N} \sum_{n=1}^N \hat{D}(n)$$

The squared magnitude of the averaged differentials is normalized with the variance of the channel estimate as

where \tilde{C}_k is the normalized average of the channel estimate over the slot.

Based on the above analysis, we have

$$\Omega_k^2 = \frac{|\tilde{D}_k|^2}{\tilde{C}_k^2},$$

which reflects the true Doppler spread. However, at this time, it is not yet known what \tilde{C}_k is.

Where

it is noted that the sub-blocks are assumed to be uncorrelated and the following holds:

$$(3) \quad \tilde{C}_k = \frac{1}{N} \sum_{n=1}^N |\hat{c}(n)|^2.$$

From the requirement of low complexity, we can take the average of N to be 1.

5 Then, all the sub-blocks over the slot are averaged to calculate the instantaneous Doppler as

$$(4) \quad \hat{f}_{D,\text{inst}} = \frac{1}{2\pi} \sqrt{\frac{2}{K} \sum_{k=1}^K \Omega_k^2}.$$

The instantaneous Doppler spread estimates over several slots are averaged using

$$(5) \quad \hat{f}_d(s) = \hat{f}_d(s-1) \cdot \beta + \hat{f}_{d,\text{inst}} \cdot (1-\beta),$$

where β is chosen as 0.99, in the exemplary embodiment. Since Doppler spread varies slowly over several slots, this is a reasonable value for β . However, it is possible to have different β values while the algorithm converges to the actual Doppler value from the initial Doppler estimate.

15 It is preferable that the instantaneous Doppler spread estimates be calculated only if the CRC checks. If the CRC does not check, the averaged Doppler spread value is not updated using the instantaneous value of the current slot, so that the instantaneous Doppler values are not used if they are not reliable. The foregoing approach to estimating Doppler spread values improves the estimation for low signal-to-noise ratios and low

20 Doppler spread values.

If the Doppler spread estimate is higher than a threshold value Th_D , a conventional demodulator is favored. If the Doppler estimate is less than the threshold value Th_D , a joint demodulator is favored.

25 Since mobile communications systems are interference limited, channel quality measurements reflect the interference level in the received signal. For the case of

continuous transmission, one contemplated, but not exclusive, technique for obtaining channel quality information $C/(I + N)$ or SINR is by using the synch word of the desired signal during the initial acquisition process. In slow fading channels, the $C/(I + N)$ estimate does not change significantly over the slot, so the synch word can be used.

The channel quality information is obtained using the synch word of the desired signal during the initial acquisition process. The received signal is modeled as

$$(6) \quad r(n) = C_d S_d + C_i S_i + \omega(n)$$

where C_d and S_d represent the channel and symbol sequence of the desired signal, C_i and S_i represent the channel and symbol sequence of the strongest interferer, and $\omega(n)$ represents the residual interference (weaker interferers) plus the thermal noise.

Using the channel estimate of the desired signal \hat{C}_d , the desired signal's synch word S_d and the received signal r over the synch word, the noise (thermal noise plus total interference) is calculated as

$$(7) \quad I_pN = K \frac{1}{K} \sum_{k=1}^K |r(k) - \hat{r}_d(k)|^2$$

where K is the length of the synch word, and $\hat{r}_d = \hat{C}_d S_d$ is the estimate of the desired signal over the synch word. The length of the synch word determines the accuracy of the noise estimate.

The residual noise indicates the level of interference (assuming interference is the only source of noise, thermal noise is negligible) in the received signal. Once the noise estimate is obtained, using the knowledge of the channel estimate, the carrier-to-interference plus noise ratio $C/(I+N)$, or signal-to-interference plus noise ratio SINR, is calculated as

$$(8) \quad \text{SINR} = \frac{C}{I_pN}$$

where L is the length of the desired channel response. In general, L is the length assumed by the conventional demodulator. Using this information, a decision can be made whether to use conventional demodulation or consider using joint demodulation.

Essentially, the residual error term, which is obtained from the best path metric values using a conventional single user demodulator, provides an estimate of noise plus interference in the received signal. In addition, the channel estimate of the desired signal, which is obtained over the synch word, provides the desired signal power estimate. From

these two estimates, the carrier-to-interference plus noise ratio SINR can be calculated. If the estimated SINR level is higher than a preselected threshold level Th_{SINR} , the conventional demodulator is used, as it performs well for these values. Otherwise, a flag is raised to consider using joint demodulation. If the joint demodulator flag is raised, a check is made for the existence of a dominant interferer.

If it is determined that the SINR estimate level is lower than the threshold level Th_{SINR} , a flag is raised that there might be a strong interference in the received signal. This does not necessarily mean that a dominant interferer is present. An SINR estimate lower than Th_{SINR} might be caused by something other than a dominant interferer, such as by a combination of many weak interferers or because the desired signal is in deep fade. Whatever the reason is, the adaptive receiver runs the joint acquisition over the synch word of the desired signal, on the assumption that a strong interferer is present. The residual noise plus interference is then calculated using the information developed about the desired signal channel estimates, the interfering signal channel estimates, the desired synch word, the interferer symbol estimates over the synch word, and the received signal. If there were a strong interference in the received signal, it is assumed that joint acquisition would take that into account and the residual noise term would be small. Therefore, the noise plus total interference term ($I + N$), I_pN , that was obtained by conventional acquisition, is compared to the noise plus residual interference term ($I_{residual} + N$), RI_pN , that was obtained by joint acquisition. If the ratio of I_pN to RI_pN is greater than a preselected threshold value $Th_{dom-int}$, i.e., if $10 * \log_{10}\left(\frac{I_pN}{RI_pN}\right) > Th_{dom-int}$, it is assumed that there is a dominant interferer, and joint demodulation is used for the rest of the slot using initial acquisition parameters obtained by joint acquisition. Otherwise, conventional demodulation is performed using the initial parameters obtained by conventional acquisition.

To determine whether a channel is dispersive or non-dispersive, a parallel approach may be used where models with a different number of taps are assumed for the synchronization and initial acquisition over the synch word. See, for example Bottomley et al. U.S. Patent application 08/897,309, filed July 21, 1997. Then, based upon the performance of both models over the synch word, a decision is made as to which model to use during demodulation.

Referring to Fig. 4, the decision of which model to use during demodulation is made based upon the residual error terms, which are obtained from the best path metric values, as used. If \hat{e}_1 and \hat{e}_2 represent the error terms that are obtained by using, for example, a one-tap model for joint demodulation 36 and a two-tap model for a conventional equalizer 38, respectively, then the error terms are calculated as

$$(9) \quad \hat{e}_1 = \sum_k |r(k) - \hat{r}_1(k)|^2 \text{ (residual error of fit)}$$

and

$$(10) \quad \hat{e}_2 = \sum_k |r(k) - \hat{r}_2(k)|^2 \text{ (residual error of fit)}$$

where $r(k)$ is the received signal, and $\hat{r}_1(k)$ and $\hat{r}_2(k)$ are the best fits to the received signal using one and two-tap channel models. Based upon the Bayesian Information Criteria (BIC), the appropriate model is selected by minimizing the quantity

$$(11) \quad Q_K = 14 \ln(\hat{e}_K) + K \ln(14)$$

where $K = 1$ for a one-tap channel model and $K = 2$ for a two-tap channel model. If $Q_1 < Q_2$, a one-tap model is used. Otherwise, if $Q_1 > Q_2$, a two-tap model is used. In other words, if $10 * \log(\frac{Q_2}{Q_1}) > .8$, in the exemplary embodiment, a two-tap model is used.

Otherwise, if $10 * \log(\frac{Q_2}{Q_1}) < .8$, in the exemplary embodiment, a one-tap model is used.

If the channel is determined to be non-dispersive, which demodulator to use depends on whether the SINR of $C/(I+N)$ is good enough, i.e., is above its threshold, to use conventional demodulation. If it is not good enough, then a determination is made whether the speed is low enough and whether a strong dominant interferer is present.

If a strong dominant interferer is present and the speed is low, then joint demodulation is used. Otherwise, conventional demodulation is used. If the channel is dispersive, joint demodulation does not perform well. However, it is not known whether the dispersion or the dominant interferer is stronger. If the dominant interferer is stronger, dispersion is

ignored and joint demodulation is performed using a one-tap desired channel model. On the other hand, if the dispersion is stronger, then interference is ignored and a conventional single user two-tap equalizer or demodulator is used. The decision is made based on speed and based on comparing the metric values obtained with a two-tap model based acquisition over the sync word and joint acquisition using one-tap desired and multiple tap interferer. It presently is contemplated that if the dominant interferer is on the order of about 3-6 dB greater than the noise level, joint demodulation is used, while if the

dominant interferer is on the order of about 3-6 dB less than the noise level, conventional demodulation is used.

In general, the three control functions (Doppler spread, dominant interference, dispersiveness) can be used individually or in various combinations to select conventional or joint demodulation.

Fig. 5 shows a contemplated scenario for use of the control functions in arriving at a decision whether to use a single user demodulator or a joint demodulator. First, single user initial acquisition is performed using conventional acquisition at a block 36. Both 1-tap and 2-tap acquisition is performed. The channel is determined to be dispersive or not at a block 38. If dispersive, single-user demodulation is performed. If not, the Doppler spread estimate for the desired signal is then compared to the threshold T_D at a decision block 40. If the Doppler spread estimate exceeds the threshold T_D , then single-user demodulation is performed at a block 42 using the channel parameters estimated by means of single-user acquisition. On the other hand, if the Doppler spread estimate does not exceed the threshold T_D , the carrier-to-interference plus noise ratio SINR is estimated and compared to the threshold T_1 at a decision block 44. If $SINR > T_1$, then single-user demodulation is performed at the block 42. Otherwise, joint acquisition is performed for the desired signal and an interferer at a block 46. Next, at a decision block 48, the noise plus total interference term ($I + N$), which is represented by M_{\min} in Fig. 5 and that was obtained by conventional acquisition, is compared with the noise plus residual interference term ($I_{\text{residual}} + N$), which is represented by M_{joint} and that was obtained by joint acquisition. If $\frac{M_{\min}}{M_{\text{joint}}} > T_2$, joint demodulation is performed at a block 50. Otherwise, single-user demodulation is performed at a block 42, preferably using the channel parameters estimated by means of joint acquisition.

A second embodiment that uses all three control functions is illustrated in the flowchart in Fig. 6. This embodiment also shows how the present invention can be used with the invention in Bottomley et al., U.S. Patent Application 08/897,309 filed July 21, 1997, which describes selecting between a 1-tap channel model single user demodulator and a 2-tap channel model single user demodulator. First, single user acquisition is performed using a 1-tap channel model in block 100. Then, the resulting $C/(I+N)$ is compared to a threshold in a decision block 102. If the threshold is exceeded, then the channel is assumed to have only 1 tap and the SINR is sufficient for 1-tap demodulation,

so single user demodulation based on a 1-tap channel model is performed in block 104. If the modulation is differential, then either coherent or differential demodulation may be performed.

If the SINR is not sufficient, then single user acquisition with a 2-tap channel model is performed in block 106. The results from the 1-tap and 2-tap acquisitions are used to determine whether the channel for the desired signal is dispersive or not in decision block 108. If the channel is decided to be not dispersive, then the Doppler spread is examined in decision block 110. If the Doppler spread is high, then single user demodulation with a 1-tap channel model is performed in block 104, as the speed would be too high to consider joint demodulation. If the Doppler spread is low, then joint acquisition is performed in block 112. In decision block 114, the $(I+N)/N$ is compared to a threshold to determine if there is a dominant interferer. If so, joint demodulation is performed in block 116. Otherwise, single user demodulation with a 1-tap channel model is performed in block 104.

If the channel is decided to be dispersive, then the SINR based on the two-tap channel model is compared to a threshold at a block 118. If it exceeds the threshold, then there is sufficient SINR for two-tap single user demodulation, which is performed in block 120. Otherwise, the Doppler spread is examined in decision block 122. If the Doppler spread is high, then single user demodulation with a 2-tap channel model is performed in block 120, as the speed would be too high to consider joint demodulation. If the Doppler spread is low, then joint acquisition is performed in block 124, similarly to as it would be performed in block 112, using a 1-tap channel model. In decision block 126, the final metrics from the 1-tap channel model and the joint acquisition are compared to determine which is more important, self ISI or interference from a dominant cochannel interferer. If the residual interference plus noise from joint acquisition is smaller than the interference plus noise after single user two-tap acquisition, then joint demodulation is performed in block 116. Otherwise, single user demodulation with a 2-tap channel model is performed in block 120.

Other combinations of the control functions, as well as combinations with other control functions, is anticipated. For example, if the channel estimation is accurate enough at high speeds, then the need to examine vehicle speed (Doppler spread) can be eliminated. Also, for the case of a base station receiver, there is a separate Doppler spread for each user. Joint demodulation may be inhibited if the Doppler spread of either the desired user or an interferer is too high. Alternatively, joint demodulation may be

performed if the Doppler spread of the interferer is low, even when the Doppler spread of the desired user is high.

While the present invention has been described with respect to joint demodulation of two users, it is also applicable to joint demodulation of three or more users.

5 Specifically, if single user demodulation SINR is not sufficient, joint demodulation of the desired and one interferer may be considered. If the SINR is still not sufficient, joint demodulation of a single user and two interferers may be considered.

It is understood that depending upon the values of expected channel conditions, the environmental conditions in which the invention is used and other differences 10 between various radio systems, the particular channel condition indicator values employed in practice of the invention may differ from system to system and often are advantageously determined empirically. Also, while the invention has been described in terms of making slot-by-slot decisions whether to use conventional or joint demodulation, the invention also contemplates combining and averaging the decisions over several slots 15 or over a prescribed time period in a non-slotted system. Such an averaging technique could be used, for example, in deciding whether the channel is dispersive, which can be used for high-speed mobiles as well as low speed mobiles.

While embodiments of the invention have been described in detail, various modifications and other embodiments thereof may be devised by one skilled in the art 20 without departing from the spirit and the scope of the invention, as defined by the appended claims.

The present invention has been described with respect to flowcharts and 25 block diagrams. It will be understood that each block of the flowchart and block diagrams can be implemented by computer program instructions. These program instructions may be provided to a processor to produce a machine, such that the instructions which execute on the processor create means for implementing the functions specified in the blocks. The computer program instructions may be executed by a processor to cause a series of operational steps to be performed by the processor to produce a computer implemented process such that the instructions which execute on the 30 processor provide steps for implementing the functions specified in the blocks.

Accordingly, the illustrations support combinations of means for performing a specified function and combinations of steps for performing the specified functions. It will also be understood that each block and combination of blocks can be implemented by special

purpose hardware-based systems which perform the specified functions or steps, or combinations of special purpose hardware and computer instructions.

CLAIMS:

1. A radio receiver for receiving a signal that includes a desired carrier, a co-channel dominant interferer and noise, said radio receiver comprising:
 - a Doppler spread estimator for generating an estimated Doppler spread value of the carrier;
 - a first estimator for generating an estimated carrier-to-dominant interferer plus noise ratio value;
 - 10 a second estimator for generating an estimated dominant interferer plus noise-to-noise ratio value; and
 - a control unit, responsive to said estimated values, for controlling said radio receiver to selectively implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer.
- 15 2. A radio receiver as in claim 1, wherein said control unit compares said estimated values to respective threshold values and controls said radio receiver to implement either single-user demodulation or joint demodulations in accordance with the results of said comparisons.
- 20 3. A radio receiver as in claim 1, wherein said control unit determines whether the carrier is dispersive, and is responsive to a determination that the carrier is dispersive to control said radio receiver to implement single-user demodulation.
- 25 4. A radio receiver as in claim 1, wherein said control unit determines whether the carrier is dispersive and whether the dominant interferer is greater or less than the dispersiveness of the carrier, and is responsive to the dominant interferer being less than the dispersiveness of the carrier to control said radio receiver to implement single-user demodulation.
- 30 5. A radio receiver as in claim 1, wherein said control unit determines whether the received signal includes multiple dominant interferers, and is responsive to a determination that the received signal includes multiple dominant interferers to control said radio receiver to implement single-user demodulation.

6. A mobile terminal used in a wireless communications system for receiving a signal that includes a desired carrier, a co-channel dominant interferer and noise, said mobile terminal comprising:

11 a Doppler spread estimator for generating an estimated Doppler spread value of
5 the carrier; and a first estimator for generating an estimated carrier-to-dominant interferer plus
noise ratio value;

16 a second estimator for generating an estimated dominant interferer plus noise-to-
noise ratio value; and

21 a control unit, responsive to said estimated values, for selectively controlling said
mobile terminal to implement either single-user demodulation of the carrier or joint
demodulation of the carrier and dominant interferer.

26 7. A mobile communications system base station for receiving a signal that
includes a desired carrier, a co-channel dominant interferer and noise, said base station
comprising: a Doppler spread estimator for generating an estimated Doppler spread value;
and a first estimator for generating an estimated carrier-to-dominant interferer plus
noise ratio value;

31 a second estimator for generating an estimated dominant interferer plus noise-to-
noise ratio value; and a control unit, responsive to said estimated values, for selectively controlling said
base station to implement either single-user demodulation of the carrier or joint
demodulation of the carrier and dominant interferer.

36 8. A radio receiver for receiving a signal that includes a desired carrier, a co-
channel dominant interferer and noise, said radio receiver comprising:

41 a control unit for controlling said radio receiver to selectively implement either
single-user demodulation of the carrier or joint demodulation of the carrier and dominant
46 interferer, the control unit determining whether the carrier is dispersive, and being
responsive to a determination that the carrier is dispersive to control said radio receiver to
implement single-user demodulation, and being responsive to a determination that the
carrier is not dispersive, for generating an estimated Doppler spread value of the carrier,

and being responsive to an estimated Doppler spread value that is at least equal to a threshold T_D , to control said radio receiver to implement single-user demodulation;

the control unit, responsive to the estimated Doppler spread value being less than the threshold T_D , for generating an estimated carrier-to-dominant interferer plus noise ratio value SINR, and being responsive to the SINR being at least equal to a threshold T_1 , to control said radio receiver to implement single-user demodulation; and

the control unit, responsive to the SINR being less than the threshold T_1 , for performing acquisition of the carrier and the dominant interferer and for generating a noise plus total interference value M_{\min} by single-user acquisition and a noise plus residual interference value M_{joint} by joint acquisition, and being responsive to the value

of the ratio $M_{\min} / M_{\text{joint}}$ being at least equal to a threshold T_2 , to control said radio receiver to implement joint demodulation and to the value of the ratio $M_{\min} / M_{\text{joint}}$ being less than said threshold T_2 , to control said radio receiver to implement single-user demodulation.

15

9. A radio receiver as in claim 8, wherein the control unit determines whether the received signal includes a single dominant interferer or multiple dominant interferers, and is responsive to a determination that there are multiple dominant interferers to control said radio receiver to implement single-user demodulation.

20

10. A radio receiver as in claim 8, wherein $\text{SINR} = C/(I + N)$, where C is the signal power of the carrier, I is the signal power of the dominant interferer, and N is the signal power of all other channel noise, including all other co-channel interferers plus thermal noise.

25

11. A mobile terminal for receiving a signal that includes a desired carrier, a co-channel dominant interferer and noise, said cellular mobile terminal comprising:
a control unit for controlling said cellular mobile terminal to selectively implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer;

12. A method for receiving a signal that includes a desired carrier, a co-channel dominant interferer and noise, comprising:

1575.1.1

the control unit determining whether the carrier is dispersive, and is responsive to a determination that the carrier is dispersive to control said mobile terminal to implement single-user demodulation; and

the control unit, responsive to a determination that the carrier is not dispersive, for generating an estimated Doppler spread value of the carrier, and is responsive to an estimated Doppler spread value that is at least equal to a threshold T_D to control said mobile terminal to implement single-user demodulation;

the control unit, responsive to the estimated Doppler spread value being less than the threshold T_D , for generating an estimated carrier-to-dominant interferer plus noise ratio value SINR, and being responsive to the SINR being at least equal to a threshold T_1 to control said mobile terminal to implement single-user demodulation; and

the control unit, responsive to the SINR being less than the threshold T_1 , for performing acquisition of the carrier and the interferer and for generating a noise plus total interference value M_{\min} by single-user acquisition and a noise plus residual interference value M_{joint} by joint acquisition, and being responsive to the value of the ratio $\frac{M_{\min}}{M_{\text{joint}}}$ being at least equal to a threshold T_2 to control said cellular mobile terminal to implement joint demodulation and to the value of the ratio $\frac{M_{\min}}{M_{\text{joint}}}$ being less than said threshold T_2 to control said mobile terminal to implement single-user demodulation.

²⁰ 12. A mobile terminal as in claim 11, wherein the control unit determines whether the received signal includes a single dominant interferer or multiple dominant interferers, and is responsive to a determination that there are multiple dominant interferers to control said mobile terminal to implement single-user demodulation.

²⁵ 13. A mobile communications system base station for receiving a signal that includes a desired carrier, a co-channel dominant interferer and noise, said cellular system base station comprising:

a control unit for controlling said base station to selectively implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer;

the control unit for determining whether the carrier is dispersive, and being responsive to a determination that the carrier is dispersive to control said base station to implement single-user demodulation;

the control unit, responsive to a determination that the carrier is not dispersive, for generating an estimated Doppler spread value, and being responsive to an estimated Doppler spread value that is at least equal to a threshold T_D to control said base station to implement single-user demodulation;

the control unit, responsive to the estimated Doppler spread value being less than the threshold T_D , for generating an estimated carrier-to-dominant interferer plus noise ratio value SINR, and being responsive to the SINR being at least equal to a threshold T_1 to control said base station to implement single-user demodulation; and

the control unit, responsive to the SINR being less than the threshold T_1 , for performing acquisition of the carrier and the interferer and for generating a noise plus total interference value M_{min} by single-user acquisition and a noise plus residual interference value M_{joint} by joint acquisition, and being responsive to the value of the ratio $\frac{M_{min}}{M_{joint}}$ being at least equal to a threshold T_2 to control said base station to

implement joint demodulation and to the value of the ratio $\frac{M_{min}}{M_{joint}}$ being less than said threshold T_2 to control said cellular system base station to implement single-user demodulation.

20

14. A mobile communications system base station as in claim 13, wherein the control unit determines whether the received signal includes a single dominant interferer or multiple dominant interferers, and is responsive to a determination that there are multiple dominant interferers to control said cellular system base station to implement single-user demodulation.

15. A radio receiver for receiving a signal that includes a desired carrier, at least one co-channel dominant interferer and noise, said radio receiver comprising:

a Doppler spread estimator for generating an estimated Doppler spread value $F_{D,av}$ of the carrier;

a first estimator for generating an estimated carrier-to-dominant interferer plus noise ratio value $C/(I+N)$; and

a second estimator for generating an estimated dominant interferer plus noise-to-noise ratio value $(I+N)/N$; and

5.0.5. a control unit for comparing said estimated values to respective threshold values and for controlling said radio receiver, in accordance with the results of said comparisons,

to implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer.

5.0.6. a despreader for despreading the received signal using the estimated Doppler shift.

10. 16. A radio receiver as in claim 15, wherein said control unit determines whether said at least one dominant interferer is a single dominant interferer or multiple dominant interferers, and is responsive to a determination that the at least one dominant interferer is multiple dominant interferers to control said radio receiver to implement single-user demodulation.

15

17. A radio receiver as in claim 15, wherein the control unit determines whether the carrier is dispersive, and is responsive to a determination that the carrier is dispersive to control said radio receiver to implement single-user demodulation.

20

18. A method of operating a radio receiver that receives a signal that includes a desired carrier, a co-channel dominant interferer and noise, comprising:

generating an estimated Doppler spread value of the carrier;

developing an estimated carrier-to-dominant interferer plus noise ratio value;

providing an estimated dominant interferer plus noise-to-noise ratio value; and

controlling the radio receiver, in accordance with the estimated values, to selectively implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer.

19. A method as in claim 18, further comprising comparing the estimated values to respective threshold values, and controlling the radio receiver, in accordance with the results of the comparisons, to implement either single-user demodulation or joint demodulation.

20. A method as in claim 18, further comprising determining whether the carrier is dispersive, and controlling the radio receiver to implement single-user demodulation in response to a determination that the carrier is dispersive.

5 20. A method as in claim 18, further comprising determining whether the carrier is dispersive and whether the dominant interferer is greater or less than the dispersiveness of the carrier, and, in response to a determination that the dominant interferer is less than the dispersiveness of the carrier, controlling the radio receiver to implement single-user demodulation.

10

22. A method as in claim 18, further comprising determining whether the received signal includes a single dominant interferer or multiple dominant interferers, and, in response to a determination that the received signal includes multiple dominant interferers, controlling the radio receiver to implement single-user demodulation.

15

23. A method of operating a mobile terminal that receives a signal that includes a desired carrier, a co-channel dominant interferer and noise, comprising:
generating an estimated Doppler spread value of the carrier;
developing an estimated carrier-to-dominant interferer plus noise ratio value;
providing an estimated dominant interferer plus noise-to-noise ratio value; and
controlling the mobile terminal, in accordance with the values of the estimated values, to selectively implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer.

25

24. A method of operating a mobile communications system base station that receives a signal that includes a desired carrier, a co-channel dominant interferer and noise, comprising:
generating an estimated Doppler spread value of the carrier;
developing an estimated carrier-to-dominant interferer plus noise ratio value;
providing an estimated dominant interferer plus noise-to-noise ratio value; and
controlling the base station, in accordance with the values of the estimated values, to selectively implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer.

25. A method of operating a radio receiver that receives a signal that includes a desired carrier, a co-channel dominant interferer and noise, comprising:

controlling the radio receiver to selectively implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer;

5 determining whether the carrier is dispersive; and, in response to a determination that the carrier is dispersive, controlling the radio receiver to implement single-user demodulation;

generating, in response to a determination that the carrier is not dispersive, an estimated Doppler spread value of the carrier, and, in response to an estimated Doppler

10 spread value that is at least equal to a threshold T_D , controlling the radio receiver to implement single-user demodulation;

developing, in response to the estimated Doppler spread value being less than the threshold T_D , an estimated carrier-to-dominant interferer plus noise ratio value SINR,

and, in response to the SINR being at least equal to a threshold T_1 , controlling the radio

15 receiver to implement single-user demodulation;

performing, in response to the SINR being less than the threshold T_1 , acquisition of the carrier and the dominant interferer; and

providing a noise plus total interference value M_{\min} by single-user acquisition and a noise plus residual interference value M_{joint} by joint acquisition; and, in response to

20 the value of the ratio $M_{\min}/M_{\text{joint}}$ being at least equal to a threshold T_2 , controlling the radio receiver to implement joint demodulation; and, in response to the value of the ratio

$M_{\min}/M_{\text{joint}}$ being less than the threshold T_2 , controlling the radio receiver to implement single-user demodulation.

each noise and interference element is gathered to become A.

25. A method as in claim 25, further comprising determining whether the received signal includes a single dominant interferer or multiple dominant interferers, and, in response to a determination that there are multiple dominant interferers, controlling the radio receiver to implement single-user demodulation.

each noise and interference element is gathered to become A.

30. A method as in claim 25, wherein $\text{SINR} = C/(I+N)$, where C is the signal power of the carrier, I is the signal power of the dominant interferer, and N is the signal noise and interference element is gathered to become A.

power of all other channel noise, including all other co-channel interferers plus thermal noise.

28. A method of operating a mobile terminal that receives a signal that includes a desired carrier, a co-channel dominant interferer and noise, comprising:

- controlling the mobile terminal to selectively implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer;
- determining whether the carrier is dispersive, and, in response to a determination that the carrier is dispersive, controlling the mobile terminal to implement single-user demodulation;
- generating, in response to a determination that the carrier is not dispersive, an estimated Doppler spread value of the carrier, and, in response to an estimated Doppler spread value that is at least equal to a threshold T_D , controlling the mobile terminal to implement single-user demodulation;
- 15 developing, in response to the estimated Doppler spread value being less than the threshold T_D , an estimated carrier-to-dominant interferer plus noise ration value SINR, and, in response to the SINR being at least equal to a threshold T_1 , controlling the mobile terminal to implement single-user demodulation;
- performing, in response to the SINR being less than the threshold T_1 , acquisition 20 of the carrier and the dominant interferer; and
- providing a noise plus total interference value M_{\min} by single-user acquisition and a noise plus residual interference value M_{joint} by joint acquisition, and, in response to the value of the ratio $M_{\min} / M_{\text{joint}}$ being at least equal to a threshold T_2 , controlling the mobile terminal to implement joint demodulation, and, in response to the value of the 25 ratio $M_{\min} / M_{\text{joint}}$ being less than the threshold T_2 , controlling the mobile terminal to implement single-user demodulation.

29. A method as in claim 28, further comprising determining whether the received signal includes a single dominant interferer or multiple dominant interferers, 30 and, in response to a determination that there are multiple dominant interferers, controlling the mobile terminal to implement single-user demodulation.

30. A method of operating a mobile communications system base station that receives a signal that includes a desired carrier, a co-channel dominant interferer and noise, comprising:

controlling the base station to selectively implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer; and, in response to determining whether the carrier is dispersive, and, in response to a determination that the carrier is dispersive, controlling the base station to implement single-user demodulation; and, in response to determining that the carrier is not dispersive, generating, in response to a determination that the carrier is not dispersive, an estimated Doppler spread value, and, in response to an estimated Doppler spread value that is at least equal to a threshold T_D , controlling the base station to implement single-user demodulation; and, in response to developing, in response to the estimated Doppler spread value being less than the threshold T_D , an estimated carrier-to-dominant interferer plus noise ration value SINR, and, in response to the SINR being at least equal to a threshold T_1 , controlling the base station to implement single-user demodulation; and, in response to performing, in response to the SINR being less than the threshold T_1 , acquisition of the carrier and the dominant interferer; and, providing a noise plus total interference value M_{\min} by single-user acquisition and a noise plus residual interference value M_{joint} by joint acquisition, and, in response to the value of the ratio $M_{\min}/M_{\text{joint}}$ being at least equal to a threshold T_2 , controlling the base station to implement joint demodulation, and, in response to the value of the ratio $M_{\min}/M_{\text{joint}}$ being less than the threshold T_2 , controlling the base station to implement single-user demodulation.

31. A method as in claim 30, further comprising determining whether the received signal includes a single dominant interferer or multiple dominant interferers, and, in response to a determination that there are multiple dominant interferers, controlling the base station to implement single-user demodulation.

30. A method as in claim 30, further comprising determining whether the received signal includes a single dominant interferer or multiple dominant interferers, and, in response to a determination that there are multiple dominant interferers, controlling the base station to implement single-user demodulation.

32. A method of operating a radio receiver that receives a signal that includes a desired carrier, at least one co-channel dominant interferer and noise, comprising:

generating an estimated Doppler spread value $F_{D,av}$ of the carrier;
developing an estimated carrier-to-dominant interferer plus noise ratio value C/(I + N);
producing an estimated dominant interferer plus noise-to-noise ratio value (I + N)/N;
comparing the estimated values to respective threshold values; and
controlling the radio receiver, in response to and in accordance with the results of said comparing step, to implement either single-user demodulation of the carrier or joint demodulation of the carrier and dominant interferer.

10

33. A method as in claim 32, further comprising determining whether the at least one dominant interferer is a single dominant interferer or multiple dominant interferers, and, in response to a determination that the at least one dominant interferer is multiple dominant interferers, controlling the radio receiver to implement single-user demodulation.

15
34. A method as in claim 32, further comprising determining whether the carrier is dispersive, and, in response to a determination that the carrier is dispersive, controlling the radio receiver to implement single-user demodulation.

20

35. A radio receiver for recovering a desired signal comprising:
a single user demodulator;
a joint demodulator; and
a control unit for selecting either said single user demodulator or said joint demodulator based on at least one of Doppler spread information, dispersion information, dominant interference information, and signal to interference plus noise information.

25

36. A radio receiver for recovering a desired signal comprising:
a single user demodulator without equalization;
a single user demodulator with equalization;
a joint demodulator; and
a control unit for selecting one of said single user demodulators or said joint demodulator based on at least one of Doppler spread information, dispersion information,

dominant interference information, and signal to interference plus noise information. channel dominant interferer and noise

The following table gives the mean and standard deviations for each of the 13 galloons.

349/2 3

... que se ha de tener en cuenta es la necesidad de que el sistema de control sea lo más simple y sencillo posible, sin embargo, no debe ser tan sencillo que no permita una adecuada ejecución de las operaciones.

so sehr wichtig, daß es sich nicht verhindern läßt, daß die Kinder im Kinde-
heitstadium ebenfalls einen großen und dichten Körperbau haben, und daß diese
einen gewissen Grad von Stärke und Kraft besitzen. Es ist daher nicht zu
verwundern, daß die Kinder im Kindesalter eine gewisse Anzahl von
gewaltigen und schweren Sünden begangen haben.

Concordia, 10 de junho de 2018, às 10 horas, na sede da Fazenda
Concordia, no distrito de Concordia, no Município de São Luís Gonzaga, no Estado do Rio Grande do Sul.

Another form of non-independence seen by the following pattern of time lags is called "anti-lagged" (inverted) "lag" lagged to end such as no need for disturbance to occur before a change in one variable can be observed.

In addition, a detailed discussion of the relationship between the two approaches is provided, and the new design is demonstrated by a numerical example. In Sect. 3, the performance of the proposed method is evaluated by comparing it with the existing methods.

1/5

FIG. 1

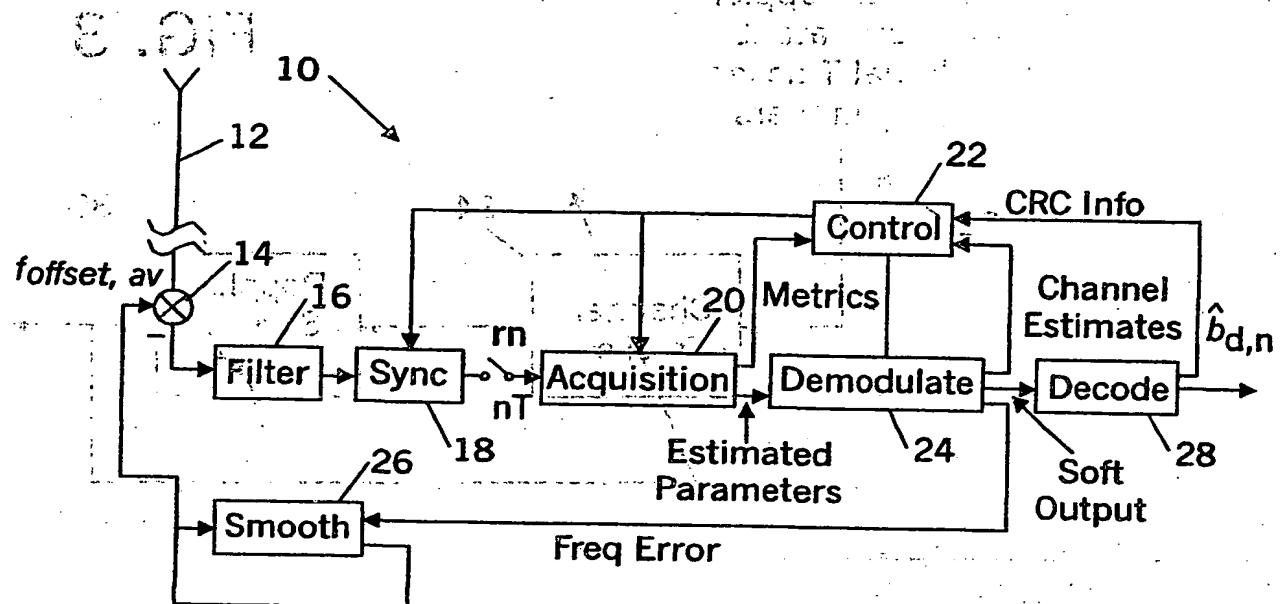
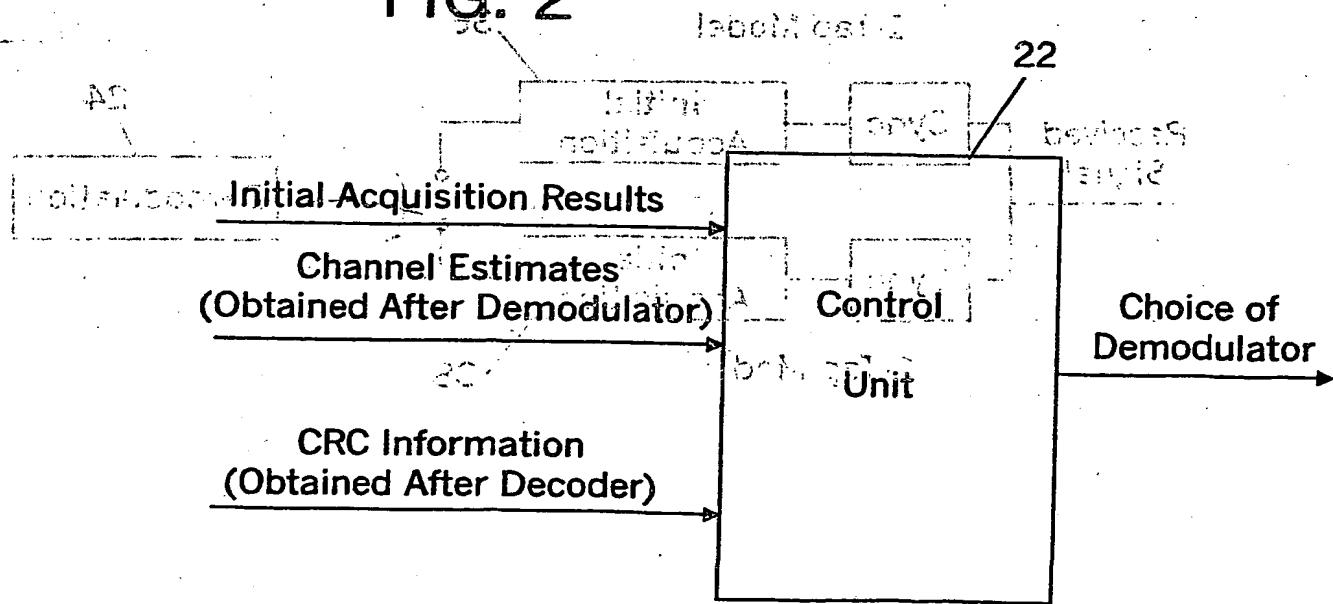


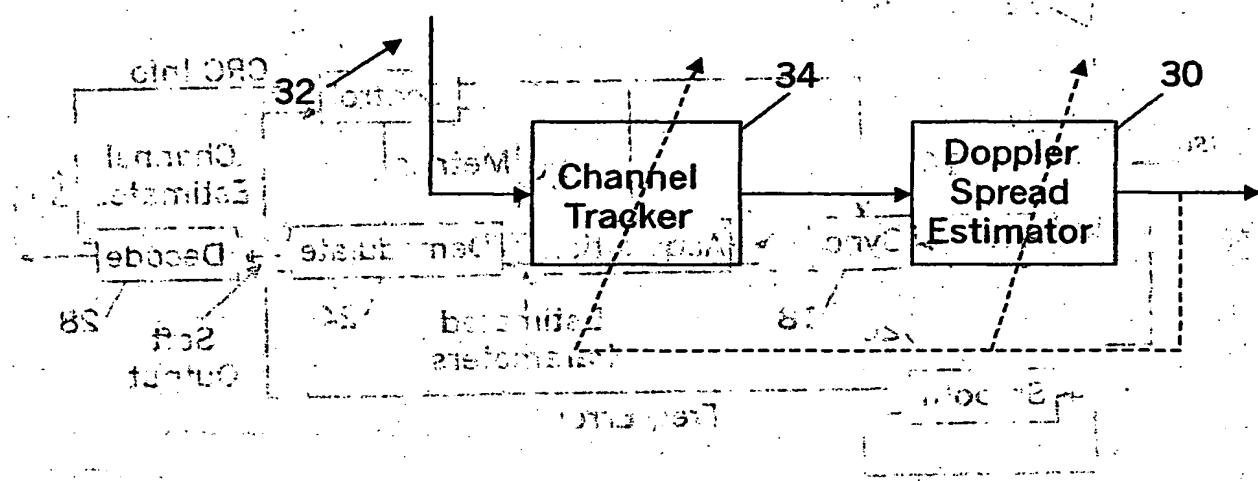
FIG. 2



2/5

S-017

**Initial Doppler
Estimate &
Initial Tracker
Parameters**

FIG. 3**FIG. 4**

1-Tap Model

Sync

Initial Acquisition

Sync

Initial Acquisition

24

Demodulation

2-Tap Model

Received Signal
(Received QPSK (16QAM) or FSK)

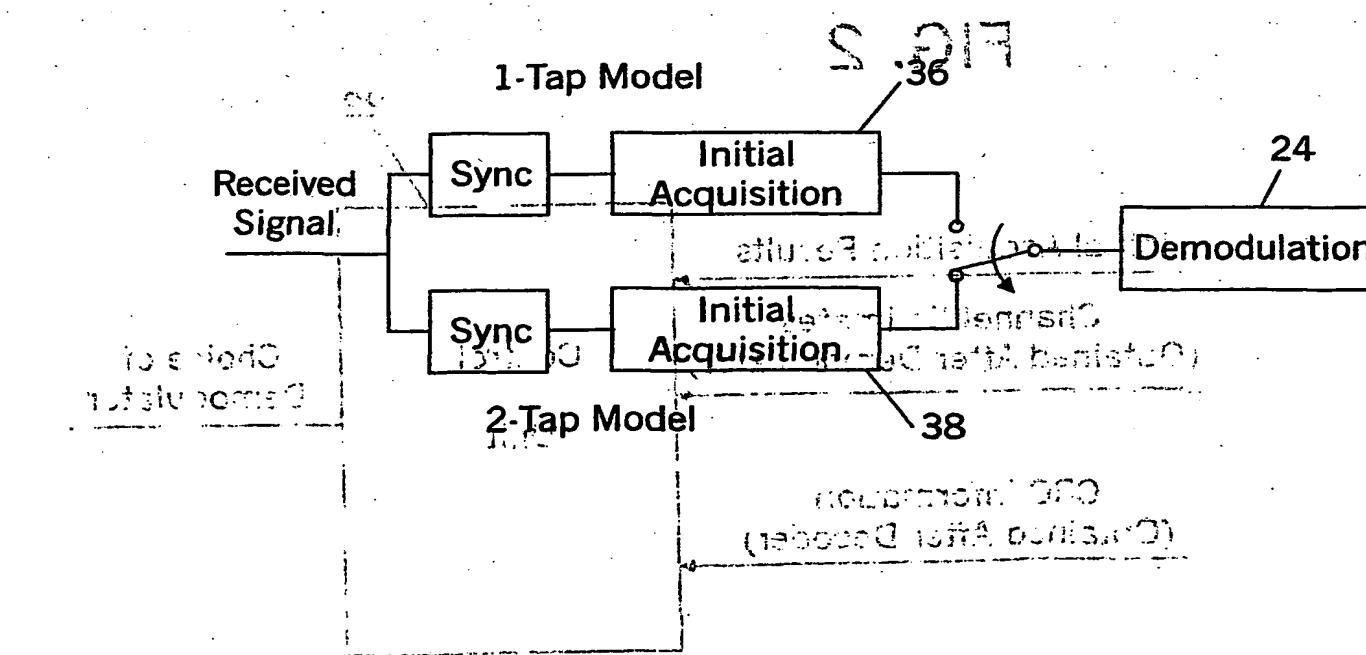
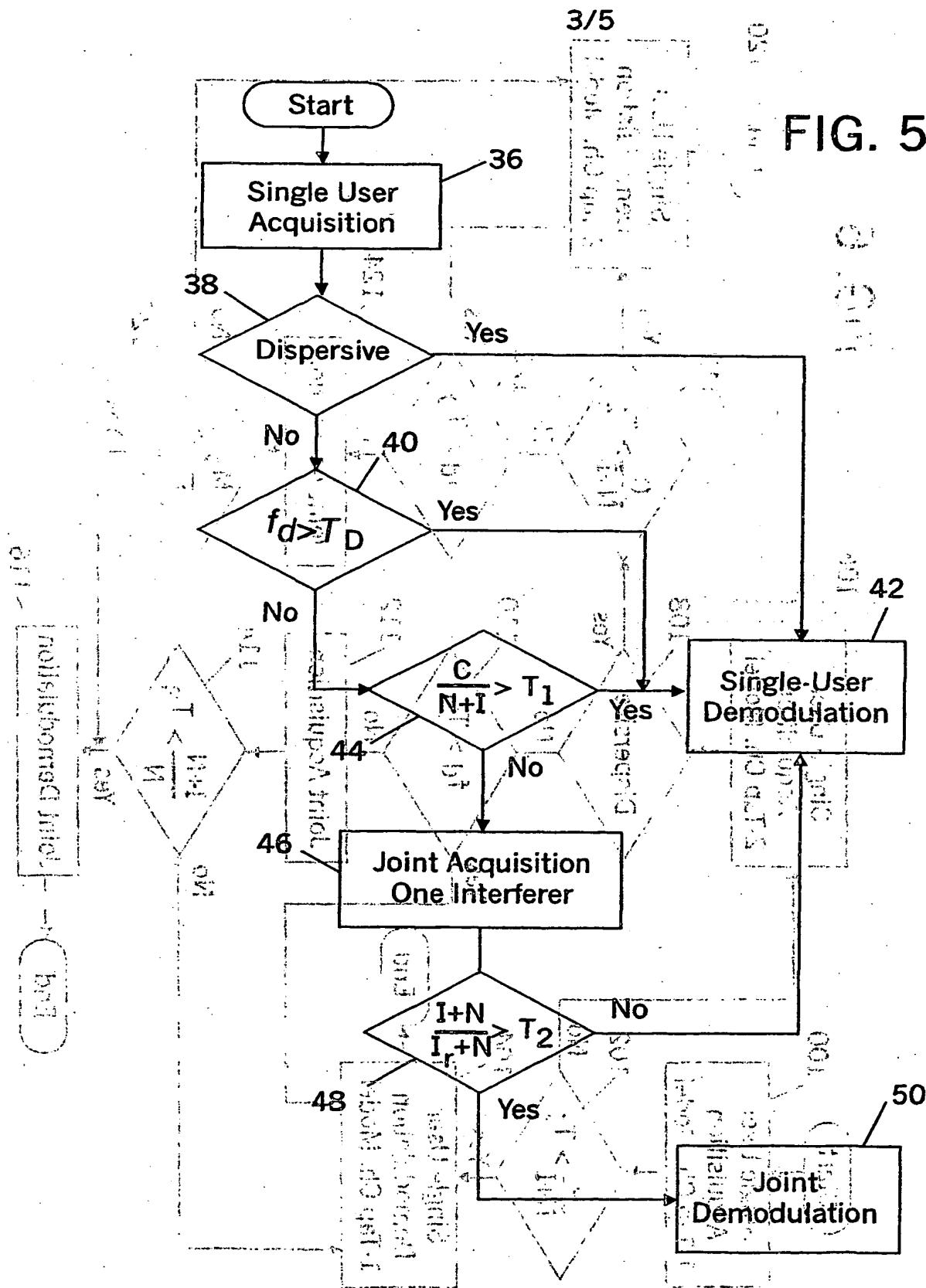
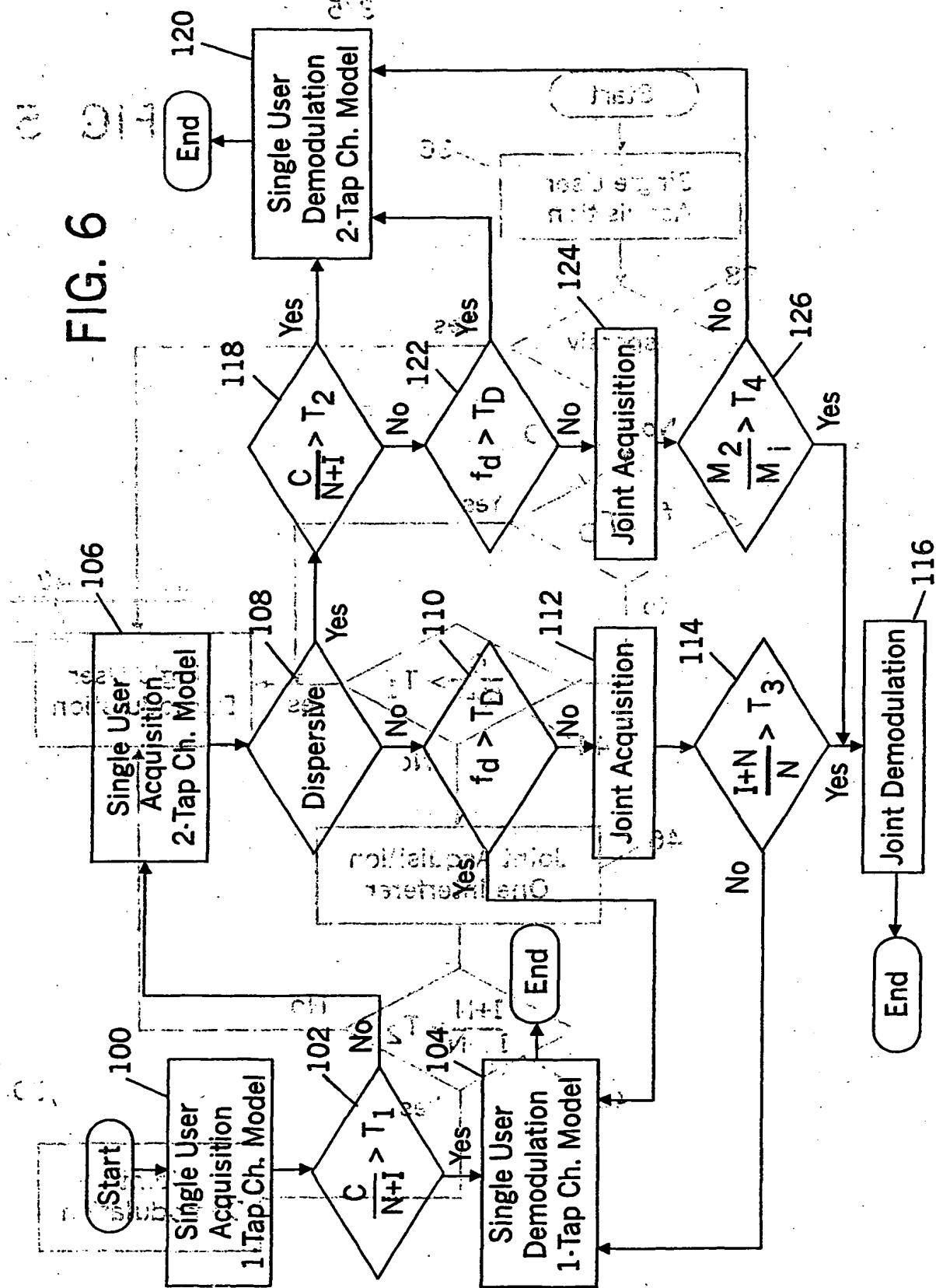


FIG. 5



4/5

FIG. 6



¹See also the discussion of the relationship between the two concepts in the section on "The Concept of Social Capital."

THE UNIVERSITY OF TORONTO LIBRARIES
UNIVERSITY OF TORONTO LIBRARY

卷之三

二三

5/5

375 *Amphibolite* - *metamorphic rocks*

1592-1593 - July 2

radiography and bone mineral analysis (RBM)

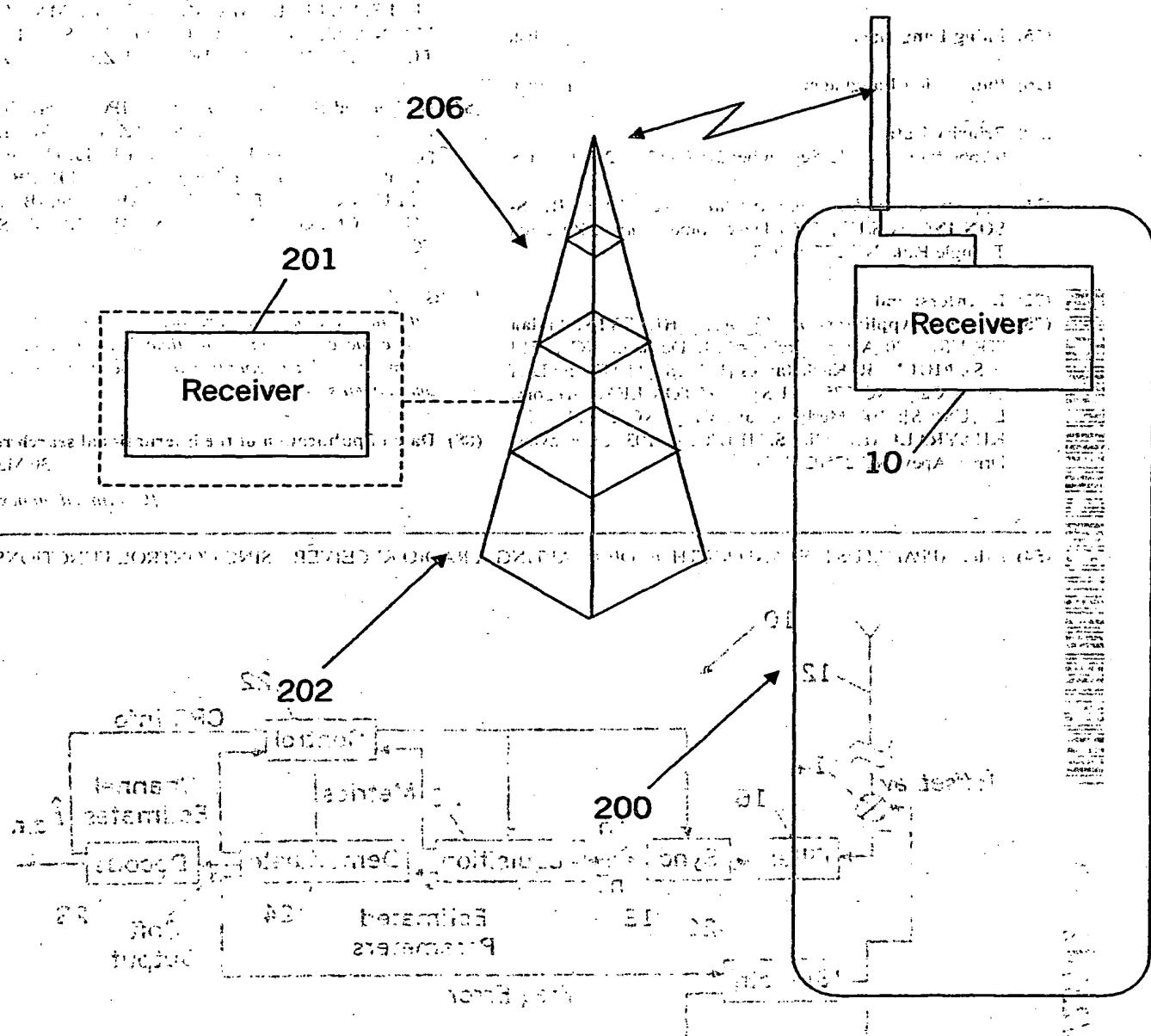
20

201

• MOGOT FORMA DE USO DA TERRA, SISTEMA DE AGRICULTURA FAMILIAR (SIAF) - 1995

卷之三

FIG. 7



PCT/CN02/00006

(12) INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(19) World Intellectual Property Organization
International Bureau



(43) International Publication Date
21 March 2002 (21.03.2002)

PCT

(10) International Publication Number
WO 02/23742 A3

(51) International Patent Classification⁷: H04L 1/00. (74) Agent: STEPHENS, Gregory, A.; Gregory A. Stephens. P.C., 301 Edgemore Avenue, Apex, NC 27502 (US).

(21) International Application Number: PCT/US01/28051 (81) Designated States (national): AE, AG, AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, BZ, CA, CH, CN, CO, CR, CU, CZ, DE, DK, DM, DZ, EE, ES, FI, GB, GD, GE, GH, GM, HR, HU, ID, IL, IN, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MA, MD, MG, MK, MN, MW, MX, MZ, NO, NZ, PL, PT, RO, RU, SD, SE, SG, SI, SK, SL, TJ, TM, TR, TT, TZ, UA, UG, US, UZ, VN, YU, ZA, ZW.

(22) International Filing Date: 7 September 2001 (07.09.2001)

(25) Filing Language: English (84) Designated States (regional): ARIPO patent (GH, GM, KE, LS, MW, MZ, SD, SL, SZ, TZ, UG, ZW), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE, TR), OAPI patent (BF, BJ, CF, CG, CI, CM, GA, GN, GQ, GW, ML, MR, NE, SN, TD, TG).

(26) Publication Language: English

(30) Priority Data: 09/660,050 12 September 2000 (12.09.2000) US

(71) Applicant (for all designated States except US): ERICSSON INC., [US/US]; 7001 Development Drive, Research Triangle Park, NC 27709 (US).

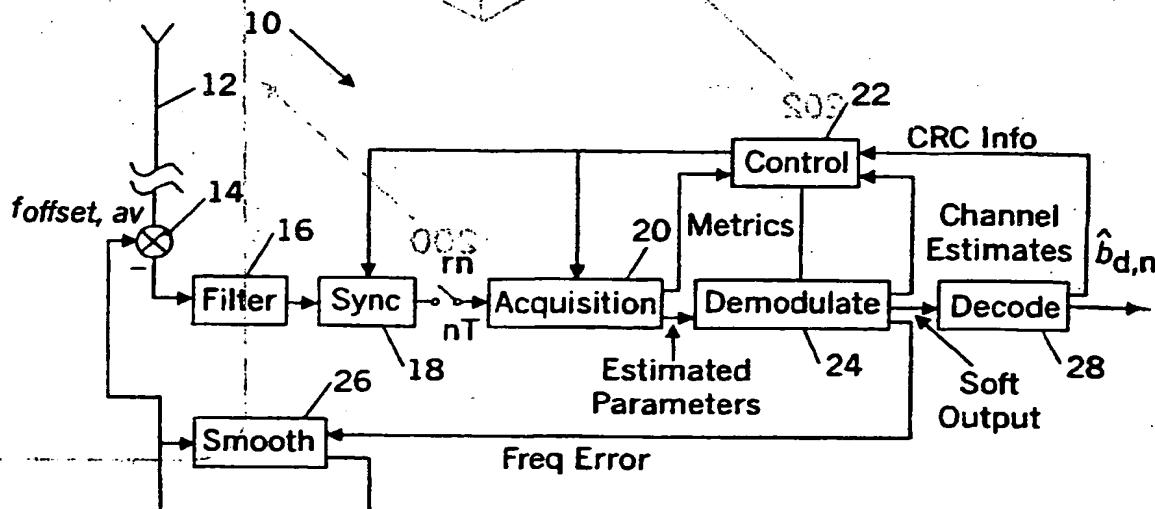
(72) Inventors; and (75) Inventors/Applicants (for US only): HUSEYIN, Arslan [TR/US]; 600 Aaron Circle, #614, Durham, NC 27713 (US). MOLNAR, Karl, James [US/US]; 110 Flying Leaf Court, Cary, NC 27513 (US). BOTTOMLEY, Gregory, E. [US/US]; 100 Merlot Court, Cary, NC 27514 (US). KHAYRALLAH, Ali, S. [LB/US]; 103 Streamview Drive, Apex, NC 27502 (US).

Published:
with international search report
before the expiration of the time limit for amending the claims and to be republished in the event of receipt of amendments

(88) Date of publication of the international search report: 30 May 2002

[Continued on next page]

(54) Title: APPARATUS FOR AND METHOD OF ADAPTING A RADIO RECEIVER USING CONTROL FUNCTIONS



WO 02/23742 A3

(57) Abstract: An adaptive radio receiver utilizes control functions derived from received signals in an interference-limited environment to decide whether to implement conventional single-user demodulation or joint demodulation. The decision to implement joint or conventional demodulation is based upon the values of estimates derived for Doppler spread, carrier-to-interference plus noise, dispersiveness and the likelihood of the presence of a dominant interferer.

SUGAR IN THE BLOOD

For two-letter codes and other abbreviations, refer to the "Guidance Notes on Codes and Abbreviations" appearing at the beginning of each regular issue of the PCT Gazette.

2000 m.s.m. - 1990 m.s.m. - 1980 m.s.m. - 1970 m.s.m. - 1960 m.s.m.

EDWARD L. BROWN - 1951 - MILITARY PERSONNEL
LAWRENCE TAYLOR - 1951 - 1952 - 1953 - 1954 - 1955
- 1956 - 1957 - 1958 - 1959 - 1960 - 1961 - 1962 - 1963

1947 DEC 7 1977-1947 8881590
8881590-09
8881590-10

6033 107-48218-6 7617 8.5 *

dated 9-16-1968 by S. Benedict, on behalf of the
US Fish & Wildlife Service, at Rockport, Maine.

On the other hand, the *luteolin* and *quercetin* groups have been shown to exert a protective effect against the development of atherosclerosis in animal models.

RECEIVED IN THE LIBRARY OF THE UNIVERSITY OF TORONTO
DECEMBER 17 1958

0223742A3 | > S 7m i anan

INTERNATIONAL SEARCH REPORT

International Application No.
PCT/US 01/28051

A. CLASSIFICATION OF SUBJECT MATTER
 IPC 7 H04L1/00 H04B1/707 H04L25/03

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)
 IPC 7 H04B H04L

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

EPO-Internal, WPI Data, PAJ, INSPEC

C. DOCUMENTS CONSIDERED TO BE RELEVANT

| Category | Citation of document, with indication, where appropriate, of the relevant passages | Relevant to claim No. |
|----------|---|-----------------------|
| X | JIUNN-TSAIR CHEN ET AL: "Low complexity joint MLSE receiver in the presence of CCI" COMMUNICATIONS, 1998. ICC 98. CONFERENCE RECORD. 1998 IEEE INTERNATIONAL CONFERENCE ON ATLANTA, GA, USA 7-11 JUNE 1998, NEW YORK, NY, USA. IEEE, US, 7 June 1998 (1998-06-07), pages 704-708, XP010284765 ISBN: 0-7803-4788-9 * 3.3 Viterbi algorithm for data retrieval. * | 35, 36 |
| A | --- | 1-34 -/- |

Further documents are listed in the continuation of box C.

Patent family members are listed in annex.

* Special categories of cited documents :

- *A* document defining the general state of the art which is not considered to be of particular relevance
- *E* earlier document but published on or after the international filing date
- *L* document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)
- *O* document referring to an oral disclosure, use, exhibition or other means
- *P* document published prior to the international filing date but later than the priority date claimed

T later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention

X document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone

Y document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art.

8 document member of the same patent family

Date of the actual completion of the international search

27 March 2002

Date of mailing of the international search report

05/04/2002

Name and mailing address of the ISA

European Patent Office, P.B. 5818 Patentlaan 2
 NL - 2280 HV Rijswijk
 Tel. (+31-70) 340-2040, Tx. 31 651 epo nl,
 Fax: (+31-70) 340-3016

Authorized officer

Martinez Martinez, V

INTERNATIONAL SEARCH REPORT

International Application No

PCT/US 01/28051

C: (Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

| Category | Citation of document, with indication, where appropriate, of the relevant passages | Relevant to claim No. |
|----------|---|--|
| A | 0001-00 REZAAIFAR E ET AL - "A HYBRID MULTI-USER 0002-00 DETECTION SCHEME FOR CDMA SYSTEMS" 0003-00 AEU INTERNATIONAL JOURNAL OF ELECTRONICS 0004-00 AND COMMUNICATIONS, GUSTAV FISHER VERLAG, 0005-01 JENA, DE, 0006-00 vol. 53, no. 2, 1999, pages 106-113, 0007-00 XP000927969 0008-00 ISSN: 1434-8411 0009-00 abstract 0010-00 page 112, right-hand column | 1, 6-8, 11, 13, 15, 23-25, 28, 30, 32, 35, 36 |
| A | WO 00 41317 A (NOKIA NETWORKS OY ; KEISALA JYRKI (FI)) 13 July 2000 (2000-07-13) abstract | 1, 6-8, 11, 13, 15, 23-25, 28, 30, 32, 35, 36 |
| A | US 5 933 768 A (HOOK MIKAEL ET AL) 3 August 1999 (1999-08-03) column 4, line 41 - line 53; figure 6 | 1, 6-8, 11, 13, 15, 23-25, 28, 30, 32, 35, 36 |
| A | US 5 375 143 A (KAZECKI HENRY L ET AL) 20 December 1994 (1994-12-20) abstract | 36 |

INTERNATIONAL SEARCH REPORT

Information on patent family members

International Application No

PCT/US 01/28051

| Patent document cited in search report | Publication date | Patent family member(s) | Publication date |
|--|------------------|---|------------------|
| WO 0041317 | A 13-07-2000 | FI 982857 A 01-07-2000 AU 3048400 A 24-07-2000 EP 1066687 A2 10-01-2001 WO 0041317 A2 13-07-2000 NO 20004315 A 30-10-2000 | |
| US 5933768 | A 03-08-1999 | AU 6640698 A 18-09-1998 CN 1249864 T 05-04-2000 EP 1012991 A1 28-06-2000 JP 2001513956 T 04-09-2001 WO 9838750 A1 03-09-1998 | |
| US 5375143 | A 20-12-1994 | US 5195106 A 16-03-1993 CA 2103392 A1 04-06-1994 GB 2273228 A 08-06-1994 CA 2049501 A1 15-05-1992 GB 2251162 A ,B 24-06-1992 US 5241545 A 31-08-1993 | |

also, a bacteriologist, a historian, a
physician, a chemist, a teacher,
and a poet.

THIS PAGE BLANK (USPTO)

THE BANK SUSPENDED

GE BLANK (USPS) *RECEIVED* *SEARCHED* *INDEXED* *FILED*

وَهُوَ الْمُنْذِرُ إِلَيْهِمْ وَهُوَ أَنْذِرَ لِلنَّاسِ
وَهُوَ أَنْذِرَ لِلنَّاسِ وَهُوَ أَنْذِرَ لِلنَّاسِ
وَهُوَ أَنْذِرَ لِلنَّاسِ وَهُوَ أَنْذِرَ لِلنَّاسِ

According to Mr. H. C. St. John, the author of "The Geology of the British Islands," the following is a list of the principal rocks of the British Isles:

I believe all these changes will be made by the end of the year. I am sure that the new system will be much more effective than the old one.

ed è stato possibile stabilire il rapporto tra la data del rientro di questi anni e l'anno delle loro nascite.

Verdades que, juntamente con las otras que quedan en el libro, al inscribirse en la memoria de los lectores, contribuirán a la formación de un pensamiento crítico y constructivo.

instead of galactose for energy, although it may also be used for synthesis of proteins.

कलेजी ने अपनी विद्यार्थी एवं शिक्षकों को इस बात का ध्यान देने की आवश्यकता समझी।

Leprosy *elutinum* *parvum* *var.* *leprosum* *var.* *leprosum*

Consequently, the following recommendations are made:

will be obtained by the method of successive approximations, and the error will be determined by the method of successive differences.

1923. It is generally agreed that the original name of the river was Tigris, and as such it is mentioned in the earliest known records, and in the legends of Gilgamesh.

THE COUNCIL OF THE STATE OF CALIFORNIA, at Sacramento, on the 1st day of January, A.D. 1850.

High - and High AD another set of results.

DATA FROM THE 1980 CENSUS OF POPULATION - 1981

**This Page is Inserted by IFW Indexing and Scanning
Operations and is not part of the Official Record**

BEST AVAILABLE IMAGES

Defective images within this document are accurate representations of the original documents submitted by the applicant.

Defects in the images include but are not limited to the items checked:

- BLACK BORDERS**
- IMAGE CUT OFF AT TOP, BOTTOM OR SIDES**
- FADED TEXT OR DRAWING**
- BLURRED OR ILLEGIBLE TEXT OR DRAWING**
- SKEWED/SLANTED IMAGES**
- COLOR OR BLACK AND WHITE PHOTOGRAPHS**
- GRAY SCALE DOCUMENTS**
- LINES OR MARKS ON ORIGINAL DOCUMENT**
- REFERENCE(S) OR EXHIBIT(S) SUBMITTED ARE POOR QUALITY**
- OTHER:** _____

IMAGES ARE BEST AVAILABLE COPY.

As rescanning these documents will not correct the image problems checked, please do not report these problems to the IFW Image Problem Mailbox.

THIS PAGE BLANK (USPTO)